EE359 – Lecture 17 Outline

- Announcements:
 - new HW posted, due Friday
 - End-of-Quarter schedule and possible bonus lecture
 - No lecture March 5
 - Advanced topics lecture; will extend last class March 12 (1-2:50 or 1:30-3:30)
 - Final exam: Tues March 17, 3:30-6:30pm, here. More details soon
- FFT implementation of MCM (OFDM)
- Implementation Challenges in OFDM
- Fading across Subcarriers
- MIMO-OFDM

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How does this reduce search complexity?

 $\hat{x} = \arg\min |Q^H y - Rx|^2$ $\hat{x} = \arg\min |y - Hx|$ $x:|Q^H y - Rx| < r$ Need to compute $|Q^h y - Rx| \forall x \in X^{M_t}$ to see if less than r

• Use tree search and upper triangular properties of R to prune search for $x=(x_1,x_2,x_3,...,x_N)$

If $|Q^h y - r_{II} x_I| > r$, prune entire branch $\begin{bmatrix} r_{22} & r_{23} \\ 0 & r_{33} \end{bmatrix} \begin{bmatrix} x_2 \\ x_3 \end{bmatrix}$

Review of Last Lecture

- MIMO RX Design (see supplemental handout):
 - Optimal Receiver is ML: finds input symbol most likely to have resulted in received vector, exponentially complex in M_t
 - Linear Receivers: First performs linear equalization: $\tilde{x} = Ay$ then quantizes \tilde{x} to nearest constellation point $x \in X^{M_t}$
 - Zero-Forcing ($A = H^{\dagger}$, the Moore-Penrose pseudo inverse of H): (if H invertible, equals inverse, else $\mathbf{H}^{\dagger} = (\mathbf{H}^H \mathbf{H})^{-1} \mathbf{H}^H$); forces offdiagonal terms to zero $(\widetilde{x}_i = x_i + \widetilde{n}_i; \widetilde{n} = \mathbf{H}^{\dagger}n$, enhances noise)
 - Minimum Mean Square Error ($A = H^H(HH^H + \lambda I)^{-1}$): $\lambda \propto 1/SNR$ Balances zero forcing against noise enhancement
- Sphere Decoder: Uses QR decomposition of H
 - Considers possibilities within sphere of transformed received symbol.
 - · If minimum distance symbol is within sphere, optimal, otherwise null is returned

$$\hat{x} = \arg\min |y - Hx|^2$$
 $\frac{Hx+n}{}$

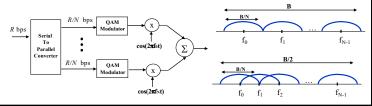


 $\hat{x} = \arg\min |Q^{H} y - Rx|^{2}$ $x:|Q^{\bar{H}}y-Rx|< r$ $Q^H y = Rx + Q^H n$

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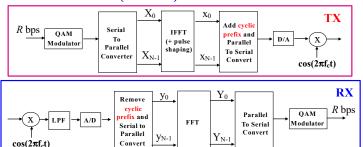
Review Continued

- ISI Mitigation: Can mitigate ISI with equalization (not covered), multicarrier modulation, or spread spectrum
- Multicarrier Modulation: breaks data into N substreams (B/N<B_c); Substreams modulated onto separate carriers
 - Substream passband BW is B/N for B total BW
 - B/N<Bc implies flat fading on each subcarrier (no ISI)
 - Can overlap channels for f_i - f_{i+1} = T_N =N/B (ortho. carriers)



FFT Implementation of MCM (OFDM)

- Use IFFT at TX to modulate symbols on each subcarrier
- Cyclic prefix makes linear convolution of channel circular, so no interference between FFT blocks in RX processing
- Reverse structure (with FFT) at receiver



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Main Points

- MCM implemented with IFFTs/FFT (OFDM)
 - Block size depends on data rate relative to delay spread
- OFDM challenges: timing/frequency offset, PAPR
- Subcarrier fading degrades OFDM performance
 - Compensate through precoding (channel inversion), coding across subcarriers, or adaptation
- OFDM naturally combined with MIMO
 - Orthogonal in space/freq; extended matrix representation
 - 4G Cellular and 802.11n/ac/ax all use OFDM+MIMO

OFDM Design Issues

- Timing/frequency offset:
 - Impacts subcarrier orthogonality; self-interference
- Peak-to-Average Power Ratio (PAPR)
 - Adding subcarrier signals creates large signal peaks
 - Solve with clipping or PAPR-optimized coding
- Different fading across subcarriers
 - Mitigate by precoding (fading inversion), adaptive modulation over frequency, and coding across subcarriers
- MIMO-OFDM
 - Apply OFDM across each spatial dimension
 - Can adapt across space, time, and frequency
 - MIMO-OFDM represented by a matrix, extends matrix representation of OFDM alone (considered in HW)

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