

Numerical Solution of a Two-disk Problem

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1 Introduction

A *multidisk problem* in frequency domain control system synthesis is a problem in which the designer wishes to find a feedback compensator that minimizes the H_∞ -norm of a certain transfer matrix related to the system being designed, subject to constraints on the H_∞ -norms of one or more other transfer matrices related to the system. These problems typically arise when a designer is asked to minimize a performance index such as the gain from a disturbance to a plant output while providing robust stabilization of the plant and/or satisfying constraints on other performance indices.

There are no known analytical solutions to general multidisk problems, although there are partial and approximate solutions [BH79, Kwa85] as well as an explicit solution to a special case [OF86]. In [TP88], Ting and Poolla propose and demonstrate a technique for finding suboptimal solutions to linear two-disk problems.

The purpose of this short paper is to show that while multidisk problems may not have analytical solutions, they can often be posed as infinite-dimensional convex programming problems, and so can be readily solved numerically. In fact, a program called *qdes*, described in [BBB*88], can be used to generate and solve finite-dimensional approximations to such infinite-dimensional programming problems. *Qdes* was used to solve the example two-disk problem considered in [TP88]; the optimal performance index was found to be over four times better than the "approximate solution" found in [TP88]. This short paper outlines how *qdes* was used to solve this example two-disk problem.

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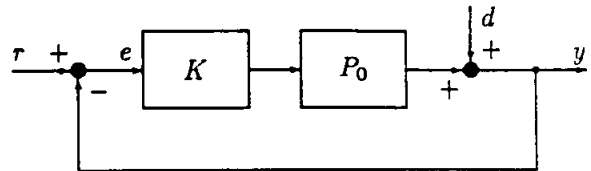


Figure 1: Block diagram for the example problem.

2 Example two-disk problem

The problem, as stated in [TP88], is

$$\inf_{K \text{ stabilizing } P_0} \|W_1(I + P_0K)^{-1}\|_\infty$$

subject to

$$\|W_2P_0K(I + P_0K)^{-1}\|_\infty \leq 1,$$

where

$$P_0(s) = \frac{s-2}{s-12}$$

$$W_1(s) = \frac{1}{30} \left(\frac{s+6}{s+1} \right)$$

$$W_2(s) = \frac{1}{3} \left(\frac{s+1}{s+2} \right) \left(\frac{(s+6)^2}{s^2+2s+37} \right).$$

The objective is to minimize the W_1 -weighted H_∞ -norm of the gain from the disturbance d to the plant output y , as shown in Figure 1, and the constraint is obtained from the specification that K stabilize all plants in the family

$$C = \{P_0(I + \Delta W_2) : \|\Delta\|_\infty < 1\}.$$

The derivation of the H_∞ constraint from this robustness specification is described in [DS81, Zam81].

It is easy to pose the problem as a convex optimization problem. In the terminology of

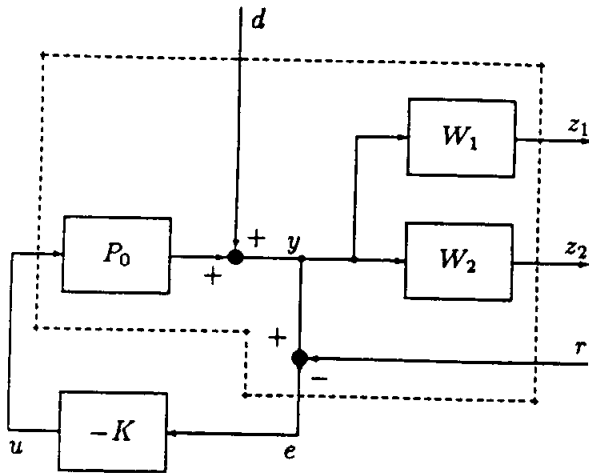


Figure 2: Problem posed in terms of H_∞ -norms of closed-loop transfer functions.

[BBB*88], u is the single *control input* and e is the single *measured variable*. As shown in Figure 2, the problem can be stated as that of finding the compensator K that stabilizes P_0 and minimizes the H_∞ -norm of the closed-loop transfer function from the *exogenous input* d to the *regulated variable* z_1 , subject to the constraint that the H_∞ -norm of the closed-loop transfer function from the exogenous input r to the regulated variable z_2 be less than or equal to one. In general any multidisk problem of the form

Minimize the H_∞ -norm of one closed-loop transfer matrix subject to upper bounds on the H_∞ -norms of other closed-loop transfer matrices

is a convex programming problem.

The solution of the problem using `qdes` proceeds as follows. A stable coprime factorization of P_0 as $P_0 = ND^{-1}$ with $YN + YD = I$, taken from [TP88], is

$$N(s) = \left(\frac{s-2}{s+6} \right); \quad D(s) = \left(\frac{s-12}{s+6} \right);$$

$$X(s) = 1.8; \quad \text{and} \quad Y(s) = -0.8.$$

The set of achievable stable closed-loop transfer matrices from the exogenous input vector $[d \ r]^T$ to the regulated variable vector $[z_1 \ z_2]^T$ can be parametrized as

$$\{T_1 + T_2QT_3 : Q \text{ stable}\}$$

where the parameter Q is a SISO transfer function and T_1 , T_2 , and T_3 are stable transfer matrices of appropriate sizes computed from the coprime factorization as shown in [BBB*88]. The compensator corresponding to a given parameter Q is

$$K = (X + DQ)(Y - NQ)^{-1}.$$

The version of `qdes` available to solve this problem is for use with discrete-time systems, so the problem is converted to an equivalent discrete-time problem by mapping the $j\omega$ -axis in the S -plane on to the unit circle in the z -plane with the substitution (bilinear transformation) $s = 20 \left(\frac{z+1}{z-1} \right)$. The solution space is made finite-dimensional by assuming that the optimal parameter $Q(z)$ is closely approximated by a finite impulse response (FIR) filter with 20 taps.

The fragment of `qdes` source code found below shows how the problem is specified to `qdes`.

```

minimize {
    max_Mag_H[Z1][D];
}

subject_to {
    max_Mag_H[Z2][R] <= 1.0;
}

```

This fragment specifies the objective and constraint; the dimensions of the problem and the values of T_1 , T_2 , and T_3 must also be specified. The output of `qdes` is a list of the FIR filter coefficients corresponding to the optimal $Q(z)$.

For the two-disk problem, `qdes` finds a 20-tap Q which satisfies the constraint. The objective function value for the approximation to the two-disk problem is 0.1864. Increasing the number of taps in the FIR filter $Q(z)$ to 40 and 80 does not significantly improve the solution; this gives some confidence that the solution based on the 20-tap $Q(z)$ is very nearly optimal. The compensator $K(z)$ obtained from $Q(z)$ is a 23rd-order stable compensator, which has a 20th-order minimal realization. The Hankel singular values of $K(z)$ are

$$1.427, 1.371, 0.821, 0.040, 0.033, \dots$$

It seems likely that the compensator model can be reduced to third-order, and this turns out to be the case. Using simple balance-and-truncate model reduction a third-order compensator was obtained that performed as well as the original

20th-order design. The application of the substitution (inverse bilinear transformation) $z = \left(\frac{s+20}{s-20}\right)$ gives the compensator

$$K(s) = -1.5345 \frac{(s-0.084796)(s^2+1.5166s+35.658)}{(s+0.86685)(s^2+1.3714s+24.576)}$$

With this compensator

$$\|W_1(I + P_0K)^{-1}\|_\infty = 0.1995$$

and

$$\|W_2P_0K(I + P_0K)^{-1}\|_\infty = 0.9929.$$

For comparison, the "approximate solution" in [TP88] has

$$\|W_1(I + P_0K)^{-1}\|_\infty = 0.7998$$

and

$$\|W_2P_0K(I + P_0K)^{-1}\|_\infty = 1.$$

3 Conclusion

While there are no known analytical techniques for solving general multidisk problems, many multidisk problems are susceptible to effective numerical solutions.

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