

EE359 Discussion Session 1

Signal Propagation Models

Jan 15, 2020

Administation

Let us know if you:

- Need project topic suggestions
- Have requests for material in discussion sessions
- Have conflicts with OH/Discussions

Outline

- 1 Signal representation and propagation models
- 2 Other models for path loss
- 3 System design considerations

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Baseband versus passband

Passband

$$s_p(t) = s_I(t) \cos(2\pi f_c t) - s_Q(t) \sin(2\pi f_c t)$$

Baseband

$$s_b(t) = s_I(t) + js_Q(t)$$

Baseband \rightarrow passband

$$s_p(t) = \operatorname{Re} \left(s_b(t) e^{j2\pi f_c t} \right)$$

Passband \rightarrow baseband (under narrowband assumption $T \gg \frac{1}{f_c}$)

$$s_I(t) \approx \frac{2}{T} \int_0^T s_p(t) \cos(2\pi f_c t) dt, \quad s_Q(t) \approx -\frac{2}{T} \int_0^T s_p(t) \sin(2\pi f_c t) dt.$$

When to use baseband and passband?

- Baseband signal convenient for DSP (digital signal processing) and mathematical analysis
- Passband signal (EM wave of frequency $\approx f_c$) is what is transmitted and received through medium

dB versus linear

- Power often measured in dB or dBm
- If P is in linear units (e.g. watts), then the same quantity in dB units is defined as

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- Power in dBm is relative to 1 mW:

$$P(\text{dBm}) = 10 \log_{10} \left(\frac{P(\text{in watts})}{1 \text{ mW}} \right)$$

Point-to-point wireless channel

Main effects

- 1 Path loss
 - ▶ Due to diffraction/finite area of antennas
 - ▶ Large scale effects (\sim Km)
- 2 Signal attenuation or shadowing
 - ▶ Due to absorption/scattering/reflection
 - ▶ Medium scale effects (\sim tens of metres)
- 3 Fading (we will revisit this later)
 - ▶ Due to phase cancellation/reinforcement due to multipath combining and doppler effects.
 - ▶ Small scale (\sim cm)

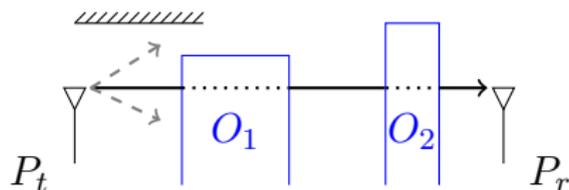


Figure: Representative wireless channel

Effect 1: Path loss

- Energy spreads due to finite area of transmitters (*diffraction*)
- Finiteness with respect to the wavelength
- Wavelength dependence is due to this diffraction effect

Free space path loss (wavelength λ)

- Line of sight: $\frac{P_r}{P_t} = \left(\frac{\lambda}{4\pi d}\right)^2 G_t G_r$ where G represent gains
- Valid in far field with no reflectors e.g. satellite communications

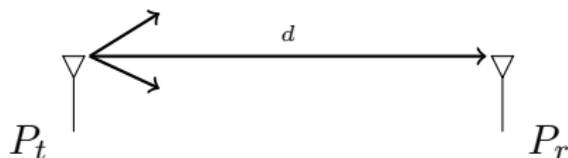


Figure: Free space path loss

Effect 2: Signal attenuation or shadowing

- Power of EM wave suffers attenuation $e^{-\alpha_i d_i}$ when it passes through object i of thickness d_i

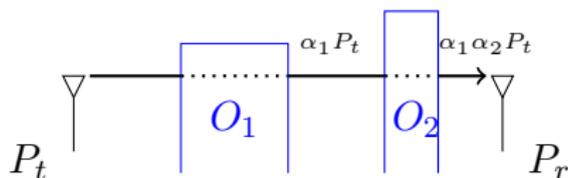


Figure: Shadowing (randomness comes from random number of objects)

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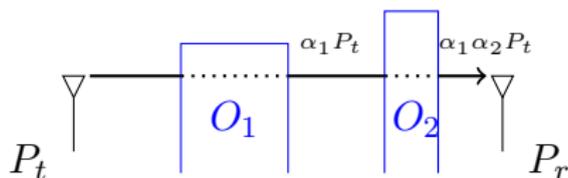


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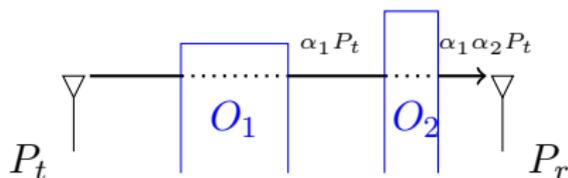


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- Account for spatial density of scatterers by modeling dependence between spatially close points, e.g.

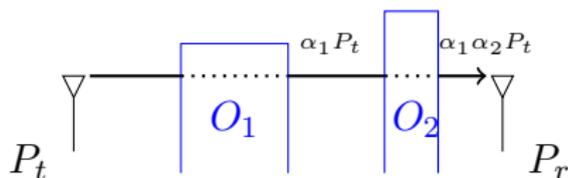


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Effect 3: Multipath combining or fading

- Multipath not always bad (if resolvable); e.g. RAKE receivers
- Variation of the order of wavelength λ since $\lambda/2$ shift causes π phase shift (thus causing cancellations $\lambda/2$ distance away from peaks)

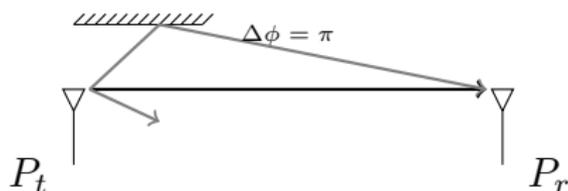


Figure: Multipath causing destructive interference

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Two ray model



Figure: Two ray model

- For small d , close to free space behavior: loss proportional to d^2 .
- For large d , $d_0 \rightarrow d_1$, $\Delta\phi \rightarrow 0$, phase inversion of reflected ray gives destructive interference: loss proportional to d^4 .
- Cutoff distance d_c capturing change in *rate of decay* can be found empirically or from heuristics (e.g. set the phase difference to be π)

Ten ray models/ray tracing

- Ten ray models
 - ▶ Models rectilinear streets with buildings on both sides
 - ▶ Transmitters and receivers close to ground
 - ▶ Includes all rays undergoing up to three reflections

- Ray-tracing models
 - ▶ Useful when geometric information available
 - ▶ Useful for site planning

Empirical/simplified/mmWave models

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 - ▶ Usually fitted to be valid over large distance and frequency ranges
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- Simplified path loss models

- ▶ Simple function of distance (far field)

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- mmWave propagation

- ▶ In addition to path loss and shadowing, faces additional attenuation from rain and atmospheric (oxygen) effects.

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Received power needs to be above threshold P_τ (noise, rate, coding, ...)

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$$P_r < P_\tau$$

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Given path loss (e.g. simplified) and shadowing models (e.g. log normal), one can compute outage probability (shadowing variance assumed to be 1)

$$P_{\text{out}} = 1 - Q\left(\frac{10 \log(\tau) - 10 \log(K P_t) + 10\gamma \log(d_0/d)}{\sigma_{\psi_{\text{dB}}}}\right)$$

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where $Q(x) \triangleq \int_x^\infty \frac{1}{\sqrt{2\pi}} e^{-y^2/2} dy$

Constant received power contour

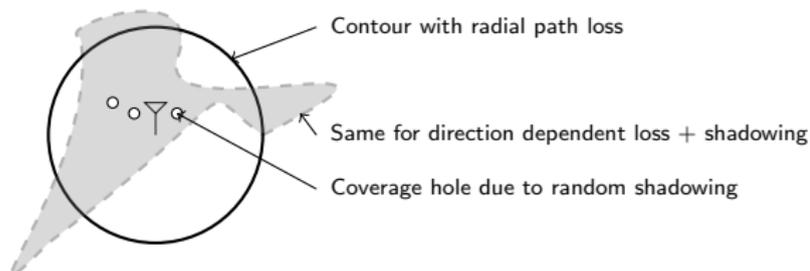


Figure: Contour plots of received power for different effects

- Without random fluctuations, contours of constant received power would be circular
- With direction dependent gains and shadowing the contours are *amoeba like*
- With random shadowing they are amoeba like with *holes*

Cell coverage area

Definition

Expected fraction of locations within a cell where the received power is above threshold τ

- If A is the area of cell

$$C = \frac{1}{A} \mathbb{E} \left[\int_A \mathbb{1}(P_r > P_\tau) dA \right]$$

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- Can be computed analytically for log normal shadowing and circular geometries
- Note that we ignore out of cell interference which would be a more pertinent limiting factor

Homework 1

Disclaimer

- Answer may vary depending on assumptions made - clearly state them
- Feel free to ignore the suggestions/references