

# AA216/CME345: PROJECTION-BASED MODEL ORDER REDUCTION

Balanced Truncation (BT)

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These slides are based on the recommended textbook: A.C. Antoulas, "Approximation of Large-Scale Dynamical Systems," Advances in Design and Control, SIAM,  
ISBN-0-89871-529-6

# Outline

- 1 Reachability and Observability
- 2 Balancing
- 3 Balanced Truncation Method
- 4 Error Analysis
- 5 Stability Analysis
- 6 Computational Complexity
- 7 Comparison with the POD Method
- 8 Application
- 9 Balanced POD Method

## └ Reachability and Observability

## └ Scope (Considered Family of Systems)

- Consider the following **stable**, high-dimensional, LTI system

$$\begin{aligned}\frac{d\mathbf{w}}{dt}(t) &= \mathbf{Aw}(t) + \mathbf{Bu}(t) \\ \mathbf{y}(t) &= \mathbf{Cw}(t) \\ \mathbf{w}(0) &= \mathbf{w}_0\end{aligned}$$

- $\mathbf{w} \in \mathbb{R}^N$ : State variables
- $\mathbf{u} \in \mathbb{R}^{in}$ : Input variables, typically  $\ll N$
- $\mathbf{y} \in \mathbb{R}^q$ : Output variables, typically  $q \ll N$

- Recall that the solution  $\mathbf{w}(t)$  of the above linear ODE can be written as

$$\mathbf{w}(t) = \phi(t, \mathbf{u}; t_0, \mathbf{w}_0) = e^{\mathbf{A}(t-t_0)} \mathbf{w}(t_0) + \int_{t_0}^t e^{\mathbf{A}(t-\tau)} \mathbf{B} \mathbf{u}(\tau) d\tau, \quad \forall t \geq t_0$$

(1)

## └ Reachability and Observability

## └ Reachability, Controllability, and Observability

## Definition

For  $T < \infty$ , a state  $\mathbf{w}(T) \in \mathbb{R}^N$  of a dynamical system is said to be **reachable** (or attainable) from an initial state  $\mathbf{w}(t_0)$  if there exists an admissible (finite energy) input function  $\mathbf{u}(\cdot)$  defined over  $[t_0, T]$  that drives the system from  $\mathbf{w}(t_0)$  to  $\mathbf{w}(T)$

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## Definition

A state  $\mathbf{w} \in \mathbb{R}^N$  of a dynamical system is said to be **controllable** to the zero state if there exists a finite-time admissible control input function  $\mathbf{u}(\cdot)$  defined over  $[t_0, T]$  ( $T < \infty$ ) that drives the system from the state  $\mathbf{w}$  to the zero state – that is,  $\phi(T, \mathbf{u}; t_0, \mathbf{w}) = \mathbf{0}_N$

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## Definition

A state  $\mathbf{w} \in \mathbb{R}^N$  of a dynamical system is said to be **unobservable** if for all  $t \geq t_0$ ,

$$\mathbf{y}(t) = \mathbf{C}\phi(t, \mathbf{0}; t_0, \mathbf{w}) = \mathbf{0}_q$$

## └ Reachability and Observability

## └ Completely Controllable Dynamical System

## Definition (R.E. Kalman, 1963)

A *linear dynamical system*  $(\mathbf{A}, \mathbf{B}, \mathbf{C})$  is said to be **completely controllable** at time  $t_0$  if it is not equivalent, for all  $t \geq t_0$ , to a system of the type

$$\begin{aligned}\frac{d\mathbf{w}^{(1)}}{dt} &= \mathbf{A}^{(1,1)}\mathbf{w}^{(1)} + \mathbf{A}^{(1,2)}\mathbf{w}^{(2)} + \mathbf{B}^{(1)}\mathbf{u} \\ \frac{d\mathbf{w}^{(2)}}{dt} &= \mathbf{A}^{(2,2)}\mathbf{w}^{(2)} \\ \mathbf{y}(t) &= \mathbf{C}^{(1)}\mathbf{w}^{(1)} + \mathbf{C}^{(2)}\mathbf{w}^{(2)}\end{aligned}$$

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- Interpretation: It is not possible to find a coordinate system in which the state variables are separated into two groups,  $\mathbf{w}^{(1)}$  and  $\mathbf{w}^{(2)}$ , such that the second group is affected neither by the first group, nor by the inputs to the system

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- This definition can be extended to linear time-variant systems

## └ Reachability and Observability

## └ Completely Observable Dynamical System

## Definition

A *linear dynamical system*  $(\mathbf{A}, \mathbf{B}, \mathbf{C})$  is said to be **completely observable** at time  $t_0$  if it is not equivalent, for all  $\underline{t \leq t_0}$ , to any system of the type

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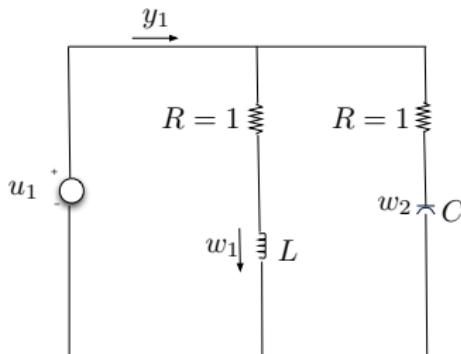
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## └ Reachability and Observability

## └ Example: Simple RLC Circuit



- For  $C = L$  and  $R = 1$ , the equation of the network shown above in terms of the current  $w_1$  flowing through the inductor and the potential  $w_2$  across the capacitor is given by

$$\begin{aligned}\frac{dw_1}{dt} &= -\frac{1}{L}w_1 + u_1 \\ \frac{dw_2}{dt} &= -\frac{1}{L}w_2 + u_1 \\ y_1 &= \frac{1}{L}w_1 - \frac{1}{L}w_2 + u_1\end{aligned}$$

## └ Reachability and Observability

## └ Example: Simple RLC Circuit

- Under the change of variable  $\bar{w}_1 = (w_1 + w_2)/2$  and  $\bar{w}_2 = (w_1 - w_2)/2$ , the previous dynamical system becomes

$$\begin{aligned}\frac{d\bar{w}_1}{dt} &= -\frac{1}{L}\bar{w}_1 + u_1 \\ \frac{d\bar{w}_2}{dt} &= -\frac{1}{L}\bar{w}_2 \\ y_1 &= \frac{2}{L}\bar{w}_2 + u_1\end{aligned}$$

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- $\bar{w}_1$  is controllable but not observable

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- $\bar{w}_1$  is controllable but not observable
- $\bar{w}_2$  is observable but not controllable
- Hence, this dynamical system is neither completely controllable nor completely observable

## └ Reachability and Observability

## └ Canonical Structure Theorem

## Theorem (Kalman, 1961)

Consider a linear dynamical system  $(\mathbf{A}, \mathbf{B}, \mathbf{C})$ . Then:

(i) There is a state space coordinate system in which the components of the state vector can be decomposed into four parts

$$\mathbf{w} = [\mathbf{w}^{(a)} \ \mathbf{w}^{(b)} \ \mathbf{w}^{(c)} \ \mathbf{w}^{(d)}]^T$$

(ii) The sizes  $N_a$ ,  $N_b$ ,  $N_c$  and  $N_d$  of these vectors do not depend on the choice of basis

(iii) The system matrices take the form

$$\mathbf{A} = \begin{bmatrix} \mathbf{A}^{(a,a)} & \mathbf{A}^{(a,b)} & \mathbf{A}^{(a,c)} & \mathbf{A}^{(a,d)} \\ 0 & \mathbf{A}^{(b,b)} & 0 & \mathbf{A}^{(b,d)} \\ 0 & 0 & \mathbf{A}^{(c,c)} & \mathbf{A}^{(c,d)} \\ 0 & 0 & 0 & \mathbf{A}^{(d,d)} \end{bmatrix}, \quad \mathbf{B} = \begin{bmatrix} \mathbf{B}^{(a)} \\ \mathbf{B}^{(b)} \\ 0 \\ 0 \end{bmatrix},$$

$$\mathbf{C} = [ \ 0 \ \ \mathbf{C}^{(b)} \ \ 0 \ \ \mathbf{C}^{(d)} \ ]$$

└ Reachability and Observability

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- The four parts of  $w$  can be interpreted as follows

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- The four parts of  $w$  can be interpreted as follows
  - part (a) is completely controllable but unobservable
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- The four parts of  $w$  can be interpreted as follows
  - part (a) is completely controllable but unobservable
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  - part (c) is uncontrollable and unobservable

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  - part (a) is completely controllable but unobservable
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  - part (a) is completely controllable but unobservable
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  - part (d) is uncontrollable and completely observable
- This theorem can be extended to linear time-variant systems

## └ Reachability and Observability

## └ Reachable and Controllable Subspaces

## Definition

The **reachable subspace**  $W_{\text{reach}} \subset \mathbb{R}^N$  of a system  $(A, B, C)$  is the set containing all reachable states of the system and

$$\mathcal{R}(A, B) = [B \ A B \ \cdots \ A^{N-1} B]$$

is the **reachability matrix** of the system

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## Definition

The **controllable subspace**  $W_{\text{contr}} \subset \mathbb{R}^N$  of a system  $(A, B, C)$  is the set containing all controllable states of the system

## └ Reachability and Observability

## └ Reachable and Controllable Subspaces

## Theorem

Given a system  $(\mathbf{A}, \mathbf{B}, \mathbf{C})$ ,

$$\mathbb{W}_{\text{reach}} = \text{Im } \mathcal{R}(\mathbf{A}, \mathbf{B})$$

## ■ Proof

- recall (1) then set  $t_0 = 0$  and  $\mathbf{w}(0) = \mathbf{0}$
- recall that  $e^{\mathbf{A}(t-\tau)} = \mathbf{I}_N + \mathbf{A}(t-\tau) + \frac{(\mathbf{A}(t-\tau))^2}{2!} + \dots$
- then  $\mathbf{w}(t) = \int_0^t \left( \mathbf{I}_N + \mathbf{A}(t-\tau) + \frac{(\mathbf{A}(t-\tau))^2}{2!} + \dots \right) \mathbf{B}\mathbf{u}(\tau) d\tau$
- for any finite  $t$ ,  $\int_0^t \frac{(t-\tau)^k}{k!} \mathbf{u}(\tau) d\tau$  acts as an in-long vector  
multiplying  $\mathbf{A}^k \mathbf{B}$  to the right  
 $\Rightarrow$  linear combination of  $\{\mathbf{B}, \mathbf{AB}, \dots, \mathbf{A}^{N-1}\mathbf{B}, \mathbf{A}^N\mathbf{B}, \mathbf{A}^{N+1}\mathbf{B}, \dots\}$
- recall Cayley-Hamilton:  $c_N \mathbf{A}^N + c_{N-1} \mathbf{A}^{N-1} + \dots + c_1 \mathbf{A} + c_0 \mathbf{I}_N = \mathbf{0}$   
 $\Rightarrow$  linear combination of the columns of  $[\mathbf{B} \ \mathbf{AB} \ \dots \ \mathbf{A}^{N-1}\mathbf{B}]$  □

## └ Reachability and Observability

## └ Reachable and Controllable Subspaces

## Corollary

- (i) If  $\mathcal{R}$  has full rank,  $\mathbf{A}\mathbb{W}_{\text{reach}} \subset \mathbb{W}_{\text{reach}}$
- (ii) The system is completely reachable if and only if  $\text{rank } \mathcal{R}(\mathbf{A}, \mathbf{B}) = N$
- (iii) Reachability is basis independent

## ■ Proof

- only the term  $\mathbf{A}^N \mathbf{B} \in \mathbb{R}^{N \times \text{in}}$  requires special attention
  - Cayley-Hamilton:  $c_N \mathbf{A}^N + c_{N-1} \mathbf{A}^{N-1} + \cdots + c_1 \mathbf{A} + c_0 \mathbf{I}_N = \mathbf{0}$

$$\Rightarrow c_N \mathbf{A}^N \mathbf{B} + c_{N-1} \mathbf{A}^{N-1} \mathbf{B} + \cdots + c_1 \mathbf{A} \mathbf{B} + c_0 \mathbf{B} = \mathbf{0}$$

$$\Rightarrow \mathbf{A}^N \mathbf{B} = -\frac{c_{N-1}}{c_N} \mathbf{A}^{N-1} \mathbf{B} - \cdots - \frac{c_1}{c_N} \mathbf{A} \mathbf{B} - \frac{c_0}{c_N} \mathbf{B}$$

□

## └ Reachability and Observability

## └ Reachability and Observability Gramians

## Definition

The **reachability (controllability) Gramian** at time  $T < \infty$  is defined as the  $N \times N$  symmetric positive semi-definite matrix

$$\mathcal{P}(T) = \int_0^T e^{\mathbf{A}\tau} \mathbf{B} \mathbf{B}^* e^{\mathbf{A}^*\tau} d\tau$$

where  $\star$  designates the transpose of the complex conjugate

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## Definition

The **observability Gramian** at time  $T < \infty$  is defined as the  $N \times N$  symmetric positive semi-definite matrix

$$\mathcal{Q}(T) = \int_0^T e^{\mathbf{A}^*\tau} \mathbf{C}^* \mathbf{C} e^{\mathbf{A}\tau} d\tau$$

## └ Reachability and Observability

## └ Reachability and Observability Gramians

## Proposition

*The columns of  $\mathcal{P}(T)$  span the reachability subspace*

$$\mathbb{W}_{\text{reach}} = \text{Im } \mathcal{R}(\mathbf{A}, \mathbf{B})$$

## └ Reachability and Observability

## └ Reachability and Observability Gramians

## Proposition

The columns of  $\mathcal{P}(T)$  span the reachability subspace

$$\mathbb{W}_{\text{reach}} = \text{Im } \mathcal{R}(\mathbf{A}, \mathbf{B})$$

## Corollary

A system  $(\mathbf{A}, \mathbf{B}, \mathbf{C})$  is reachable if and only if  $\mathcal{P}(T)$  is Symmetric Positive Definite (SPD) at some time  $T > 0$

## ■ Proof

- if  $\mathcal{P}(T)$  is SPD, define the input  $\mathbf{u}(t) = \mathbf{B}^* e^{\mathbf{A}^*(T-t)} \mathcal{P}^{-1} \mathbf{w}$ ,  $t \in [0, T]$
- starting from  $\mathbf{w}(0) = \mathbf{0}$ , the resulting final state is  $\mathbf{w}(t) = \int_0^T e^{\mathbf{A}(T-\tau)} \mathbf{B} \mathbf{u}(\tau) d\tau = \left( \int_0^T e^{\mathbf{A}(T-\tau)} \mathbf{B} \mathbf{B}^* e^{\mathbf{A}^*(T-\tau)} d\tau \right) \mathcal{P}^{-1} \mathbf{w}(T)$   
 $= \underbrace{\left( \int_0^T e^{\mathbf{A}\tau} \mathbf{B} \mathbf{B}^* e^{\mathbf{A}^*\tau} d\tau \right)}_{\mathcal{P}} \mathcal{P}^{-1} \mathbf{w}(T) = \mathbf{w}(T)$

## └ Reachability and Observability

## └ Equivalence Between Reachability and Controllability

## Theorem

*For continuous linear dynamical systems, the notions of controllability and reachability are equivalent – that is,*

$$\mathbb{W}_{reach} = \mathbb{W}_{contr}$$

## └ Reachability and Observability

## └ Unobservability Subspace

## Definition

The **unobservability subspace**  $\mathbb{W}_{\text{unobs}} \subset \mathbb{R}^N$  is the set of all unobservable states of the system and the matrix

$$\mathcal{O}(\mathbf{C}, \mathbf{A}) = [\mathbf{C}^* \ \mathbf{A}^* \mathbf{C}^* \ \cdots \ (\mathbf{A}^*)^i \mathbf{C}^* \ \cdots]^*$$

is the **observability matrix** of the system

## └ Reachability and Observability

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## Theorem

Given a system  $(\mathbf{A}, \mathbf{B}, \mathbf{C})$ ,

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## └ Reachability and Observability

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## Corollary

(i)  $\mathbf{A}\mathbb{W}_{unobs} \subset \mathbb{W}_{unobs}$

(ii) The system is completely observable if and only if  $\text{rank } \mathcal{O}(\mathbf{C}, \mathbf{A}) = N$

(iii) Observability is basis independent

## └ Reachability and Observability

## └ Infinite Gramians

## Definition

The **infinite reachability (controllability) Gramian** is defined for a **stable** LTI system as the  $N \times N$  symmetric positive semi-definite matrix

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## └ Reachability and Observability

## └ Energetic Interpretation

- $\mathcal{P}$  and  $\mathcal{Q}$  are respective bases for the reachable and observable subspaces
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- For a reachable state, the inner product based on  $\mathcal{P}^{-1}$  characterizes the minimal energy required to steer the state from  $\mathbf{0}$  to  $\mathbf{w}$  as  $t \rightarrow \infty$

$$\|\mathbf{w}\|_{\mathcal{P}^{-1}}^2 = \mathbf{w}^T \underbrace{\mathcal{P}^{-1} \mathbf{w}}_{\substack{\text{homogeneous} \\ \text{to an input}}} \quad \left( \leq \int_0^t (\mathbf{B} \mathbf{u}(\tau))^* \mathbf{B} \mathbf{u}(\tau) d\tau \right)$$

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- The inner product based on  $\mathcal{Q}$  indicates the maximal energy produced by observing the output of the system corresponding to an initial state  $\mathbf{w}_0$  when no input is applied

$$\|\mathbf{w}\|_{\mathcal{Q}}^2 = \mathbf{w}^T \mathcal{Q} \mathbf{w}$$

## └ Balancing

## └ Model Order Reduction Based on Balancing

- If  $\mathcal{P}$  is large in some direction  $\mathbf{w}$ ,  $\mathbf{w}^T \mathcal{P}^{-1} \mathbf{w}$  is small,  $\mathbf{w}$  can be reached using a small control energy, **but  $\mathbf{w}^T \mathcal{P} \mathbf{w}$  is large**
- If  $\mathcal{Q}$  is large in some direction  $\mathbf{w}$ ,  $\mathbf{w}^T \mathcal{Q} \mathbf{w}$  is large and that direction produces a large observation energy
- PMOR strategy: **Eliminate** the states  $\mathbf{w}$  that are simultaneously
  - **difficult to reach**, i.e., require a large energy  $\|\mathbf{w}\|_{\mathcal{P}^{-1}}^2$  to be reached
  - **difficult to observe**, i.e., produce a small observation energy  $\|\mathbf{w}\|_{\mathcal{Q}}^2$
- The above notions are **basis-dependent**
- One would like to consider a basis where both energy measures are **equal** or **balanced** – specifically, a basis where  $\mathbf{w}^T \mathcal{P} \mathbf{w}$  and  $\mathbf{w}^T \mathcal{Q} \mathbf{w}$  are balanced (see first two bullets)

## └ Balancing

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## └ Balancing

## └ Balancing Transformation

- Balancing transformations  $\mathbf{T}_{\text{bal}}$  and  $\mathbf{T}_{\text{bal}}^{-1}$  can be computed as follows
  - 1 compute the Cholesky factorization  $\mathcal{P} = \mathbf{U}\mathbf{U}^*$
  - 2 compute the eigenvalue decomposition of  $\mathbf{U}^* \mathcal{Q} \mathbf{U}$

$$\mathbf{U}^* \mathcal{Q} \mathbf{U} = \mathbf{K} \boldsymbol{\Sigma}^2 \mathbf{K}^*$$

where the entries in  $\boldsymbol{\Sigma}$  are ordered decreasingly

- 3 compute the transformations

$$\mathbf{T}_{\text{bal}} = \boldsymbol{\Sigma}^{\frac{1}{2}} \mathbf{K}^* \mathbf{U}^{-1}$$

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$$\begin{aligned}\mathbf{T}_{\text{bal}} &= \boldsymbol{\Sigma}^{\frac{1}{2}} \mathbf{K}^* \mathbf{U}^{-1} \\ \mathbf{T}_{\text{bal}}^{-1} &= \mathbf{U} \mathbf{K} \boldsymbol{\Sigma}^{-\frac{1}{2}}\end{aligned}$$

- Then, one can check that balancing is achieved

$$\mathbf{T}_{\text{bal}} \mathcal{P} \mathbf{T}_{\text{bal}}^* = \mathbf{T}_{\text{bal}}^* \mathcal{Q} \mathbf{T}_{\text{bal}}^{-1} = \boldsymbol{\Sigma}$$

## Definition (Hankel Singular Values)

$\boldsymbol{\Sigma} = \text{diag}(\sigma_1, \dots, \sigma_N)$  contains the **N Hankel singular values of the system** (a Hankel singular value is computed from the Hankel operator or the product of Gramian matrices ( $\mathcal{P}\mathcal{Q}$ ) associated with a LTI system and measures the energy of a corresponding internal state)

## └ Balancing

## └ Variational Interpretation

- Computing the balancing transformation  $\mathbf{T}_{\text{bal}}$  is equivalent to minimizing the following function

$$\min_{\mathbf{T} \in \text{GL}(N)} f(\mathbf{T}) = \min_{\mathbf{T} \in \text{GL}(N)} \text{trace}(\mathbf{T} \mathcal{P} \mathbf{T}^* + \mathbf{T}^{*-1} \mathcal{Q} \mathbf{T}^{-1})$$

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- For the optimal transformation  $\mathbf{T}_{\text{bal}}$ , the function takes the value

$$f(\mathbf{T}_{\text{bal}}) = 2\text{tr}(\mathbf{\Sigma}) = 2 \sum_{i=1}^N \sigma_i$$

where  $\{\sigma_i\}_{i=1}^N$  are the Hankel singular values

## └ Balanced Truncation Method

## └ Block Partitioning of the System

- Applying the change of variable  $\bar{\mathbf{w}} = \mathbf{T}_{\text{bal}}\mathbf{w}$  transforms the given dynamical system into  $(\bar{\mathbf{A}}, \bar{\mathbf{B}}, \bar{\mathbf{C}})$  where

$$\bar{\mathbf{A}} = \mathbf{T}_{\text{bal}} \mathbf{A} \mathbf{T}_{\text{bal}}^{-1}, \quad \bar{\mathbf{B}} = \mathbf{T}_{\text{bal}} \mathbf{B}, \quad \bar{\mathbf{C}} = \mathbf{C} \mathbf{T}_{\text{bal}}^{-1}$$

- Let  $1 \leq k \leq N$ ; the system can be partitioned in blocks as

$$\bar{\mathbf{A}} = \begin{bmatrix} \bar{\mathbf{A}}_{11} & \bar{\mathbf{A}}_{12} \\ \bar{\mathbf{A}}_{21} & \bar{\mathbf{A}}_{22} \end{bmatrix}, \quad \bar{\mathbf{B}} = \begin{bmatrix} \bar{\mathbf{B}}_1 \\ \bar{\mathbf{B}}_2 \end{bmatrix}, \quad \bar{\mathbf{C}} = \begin{bmatrix} \bar{\mathbf{C}}_1 & \bar{\mathbf{C}}_2 \end{bmatrix}$$

- The subscripts 1 and 2 denote the dimensions  $k$  and  $N - k$ , respectively

- The blocks with the subscript 1 correspond to the most observable and reachable states

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$$(\mathbf{A}_r, \mathbf{B}_r, \mathbf{C}_r) = (\bar{\mathbf{A}}_{11}, \bar{\mathbf{B}}_1, \bar{\mathbf{C}}_1) \in \mathbb{R}^{k \times k} \times \mathbb{R}^{k \times \text{in}} \times \mathbb{R}^{q \times k}$$

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- The **left and right** ROBs are

$$\mathbf{W} = \mathbf{T}_{\text{bal}}^*(:, 1:k) \quad \text{and} \quad \mathbf{V} = \mathbf{S}_{\text{bal}}(:, 1:k), \text{ respectively,}$$

where  $\mathbf{S}_{\text{bal}} = \mathbf{T}_{\text{bal}}^{-1}$

## └ Error Analysis

## └ Error Criterion

Definition (The Hardy space  $\mathcal{H}_\infty$ )

The  $\mathcal{H}_\infty$ -norm associated with a system characterized by a *transfer function*  $\mathbf{G}(\cdot)$  is defined as

$$\|\mathbf{G}\|_{\mathcal{H}_\infty} = \sup_{z \in \mathbb{C}_+} \|\mathbf{G}(z)\|_\infty = \sup_{z \in \mathbb{C}_+} \sigma_{\max}(\mathbf{G}(z))$$

where  $z \in \mathbb{C}_+$  if  $z \in \mathbb{C}$  and  $\Im(z) > 0$ .

## Proposition

$$(i) \quad \|\mathbf{G}\|_{\mathcal{H}_\infty} = \sup_{\omega \in \mathbb{R}} \|\mathbf{G}(i\omega)\|_\infty = \sup_{\omega \in \mathbb{R}} \sigma_{\max}(\mathbf{G}(i\omega))$$

$$(ii) \quad \|\mathbf{G}\|_{\mathcal{H}_\infty} = \sup_{\mathbf{u} \neq 0} \frac{\|\mathbf{y}(\cdot)\|_2}{\|\mathbf{u}(\cdot)\|_2} = \sup_{\mathbf{u} \neq 0} \sqrt{\frac{\int_0^\infty \|\mathbf{y}(t)\|_2^2 dt}{\int_0^\infty \|\mathbf{u}(t)\|_2^2 dt}}$$

- The  $\mathcal{H}_\infty$  norm of the error between the HDM- and PROM-based solutions will be used as an error criterion

## └ Error Analysis

## └ Theorem

## Theorem (Error Bounds)

The BT procedure yields the following error bound for the output of interest.

Let  $\{\bar{\sigma}_i\}_{i=1}^{N_{SV}} \subseteq \{\sigma_i\}_{i=1}^N$  denote the **distinct** Hankel singular values of the system and  $\{\bar{\sigma}_i\}_{i=N_k+1}^{N_{SV}}$  the ones that have been truncated. Then

$$\|\mathbf{y}(\cdot) - \mathbf{y}_r(\cdot)\|_2 \leq 2 \sum_{i=N_k+1}^{N_{SV}} \bar{\sigma}_i \|\mathbf{u}(\cdot)\|_2$$

Equivalently, the above result can be written in terms of the  $\mathcal{H}_\infty$ -norm of the system error as follows

$$\|\mathbf{G}(\cdot) - \mathbf{G}_r(\cdot)\|_{\mathcal{H}_\infty} \leq 2 \sum_{i=N_k+1}^{N_{SV}} \bar{\sigma}_i$$

where  $\mathbf{G}$  and  $\mathbf{G}_r$  are the full- and reduced-order transfer functions. Equality holds when  $\bar{\sigma}_{N_k+1} = \bar{\sigma}_{N_{SV}}$  (all truncated singular values are equal).

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**Proof.** The proof proceeds in 3 steps:

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$$\|\mathbf{E}(\cdot)\|_{\mathcal{H}_\infty} = 2(N_{\text{SV}} - N_k) \sigma$$

- 3 Use this result to show that in the general case

$$\|\mathbf{E}(\cdot)\|_{\mathcal{H}_\infty} \leq 2 \sum_{i=N_k+1}^{N_{\text{SV}}} \bar{\sigma}_i$$

## └ Stability Analysis

## └ Theorem

## Theorem (Stability Preservation)

Consider  $(\mathbf{A}_r, \mathbf{B}_r, \mathbf{C}_r) = (\bar{\mathbf{A}}_{11}, \bar{\mathbf{B}}_1, \bar{\mathbf{C}}_1)$ , a PROM obtained by BT. Then

- (i)  $\mathbf{A}_r = \bar{\mathbf{A}}_{11}$  has no eigenvalues in the open right half plane
- (ii) Furthermore, if the systems  $(\bar{\mathbf{A}}_{11}, \bar{\mathbf{B}}_1, \bar{\mathbf{C}}_1)$  and  $(\bar{\mathbf{A}}_{22}, \bar{\mathbf{B}}_2, \bar{\mathbf{C}}_2)$  have no Hankel singular values in common,  $\mathbf{A}_r$  has no eigenvalues on the imaginary axis

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## └ Numerical Methods

- Because of numerical stability issues, computing the transformations

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$$\mathbf{U}^* \mathbf{Z} = \mathbf{W} \boldsymbol{\Sigma} \mathbf{V}^*$$

- construct the matrices

$$\mathbf{T}_{\text{bal}} = \boldsymbol{\Sigma}^{-\frac{1}{2}} \mathbf{V}^* \mathbf{Z}^* \quad \text{and} \quad \mathbf{T}_{\text{bal}}^{-1} = \mathbf{U} \mathbf{W} \boldsymbol{\Sigma}^{-\frac{1}{2}}$$

## └ Computational Complexity

## └ Numerical Methods

- Because of numerical stability issues, computing the transformations

$$\mathbf{T}_{\text{bal}} = \boldsymbol{\Sigma}^{\frac{1}{2}} \mathbf{K}^* \mathbf{U}^{-1}, \quad \mathbf{T}_{\text{bal}}^{-1} = \mathbf{U} \mathbf{K} \boldsymbol{\Sigma}^{-\frac{1}{2}}$$

is usually ill-advised (computation of inverses)

- Instead, it is better advised to use the following procedure
  - start from the Cholesky decompositions of the Gramians

$$\mathcal{P} = \mathbf{U} \mathbf{U}^* \quad \text{and} \quad \mathcal{Q} = \mathbf{Z} \mathbf{Z}^*$$

- compute the SVD

$$\mathbf{U}^* \mathbf{Z} = \mathbf{W} \boldsymbol{\Sigma} \mathbf{V}^*$$

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$$\mathbf{T}_{\text{bal}} = \boldsymbol{\Sigma}^{-\frac{1}{2}} \mathbf{V}^* \mathbf{Z}^* \quad \text{and} \quad \mathbf{T}_{\text{bal}}^{-1} = \mathbf{U} \mathbf{W} \boldsymbol{\Sigma}^{-\frac{1}{2}}$$

- Proof: Recall that  $\boldsymbol{\Sigma}$  is always real-valued then compute  $\mathbf{T}_{\text{bal}} \mathcal{P} \mathbf{T}_{\text{bal}}^*$  and  $\mathbf{T}_{\text{bal}}^* \mathcal{Q} \mathbf{T}_{\text{bal}}^{-1}$  using the above SVD

## └ Computational Complexity

## └ Cost and Limitations

- BT leads to PROMs with quality and stability guarantees; however
  - the computation of a Gramian is intensive as it requires the solution of a Lyapunov equation ( $\mathcal{O}(N^3)$  operations)
  - for this reason, BT is in general impractical for large systems – say  $N \gtrsim 10^5$  (but monitor progress in the literature if interested)

## └ Comparison with the POD Method

## └ POD

Recall the theorem underlying the construction of a POD basis

**Theorem**

Let  $\hat{\mathbf{K}} \in \mathbb{R}^{N \times N}$  be the real symmetric semi-definite positive matrix defined as

$$\hat{\mathbf{K}} = \int_0^T \mathbf{w}(t) \mathbf{w}(t)^T dt$$

Let  $\hat{\lambda}_1 \geq \hat{\lambda}_2 \geq \dots \geq \hat{\lambda}_N \geq 0$  denote the ordered eigenvalues of  $\hat{\mathbf{K}}$  and let  $\hat{\phi}_i \in \mathbb{R}^N$ ,  $i = 1, \dots, N$ , denote the associated eigenvectors

$$\hat{\mathbf{K}} \hat{\phi}_i = \hat{\lambda}_i \hat{\phi}_i, \quad i = 1, \dots, N.$$

The subspace  $\hat{\mathcal{V}} = \text{range}(\hat{\mathbf{V}})$  of dimension  $k$  minimizing  $J(\Pi_{\mathcal{V}}, \mathbf{v})$  is the invariant subspace of  $\hat{\mathbf{K}}$  associated with the eigenvalues  $\hat{\lambda}_1, \dots, \hat{\lambda}_k$

- └ Comparison with the POD Method
  - └ POD for an Impulse Response

- The response of an LTI system to a single impulse input with a zero initial condition is

$$\mathbf{w}(t) = e^{\mathbf{A}t} \mathbf{B}$$

- Comparison with the POD Method
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- Consequently, the reachability Gramian is

$$\mathcal{P} = \int_0^{\mathcal{T}} e^{\mathbf{A}t} \mathbf{B} \mathbf{B}^T e^{\mathbf{A}^T t} dt = \int_0^{\mathcal{T}} \mathbf{w}(t) \mathbf{w}(t)^T dt = \hat{\mathbf{K}}$$

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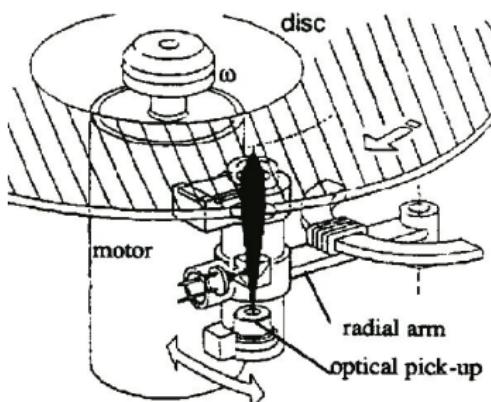
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- Unlike the BT method, the POD method does not take into account the observability Gramian to determine the PROM: therefore, every observable state may be truncated

## Application

## CD Player System (B. Salimbahrami and B. Lohmann, 2003)



- Objective: model the position of the lens controlled by a swing arm
- System with  $in = 2$  inputs
  - control voltage commanding the lens to move up and down to maintain the laser beam's focus on the disc's information layer
  - control voltage commanding the entire swing arm to move radially to keep the laser spot precisely on the data track

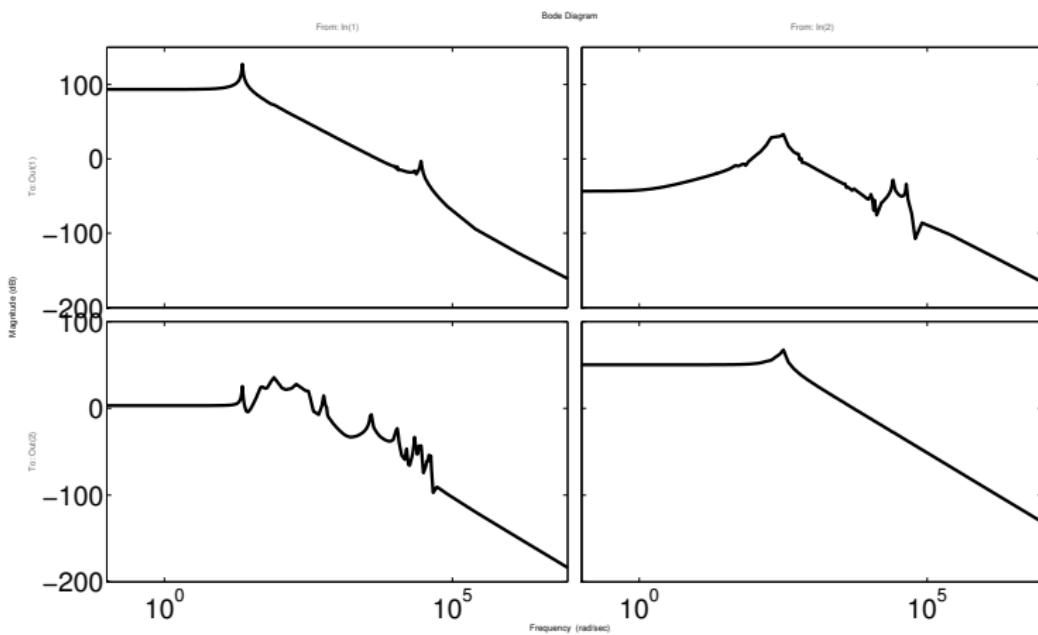
and  $q = 2$  outputs

- focus error signal (degree and direction of vertical misalignment)
- tracking error signal (degree and direction of radial misalignment)

## Application

## CD Player System (B. Salimbahrami and B. Lohmann, 2003)

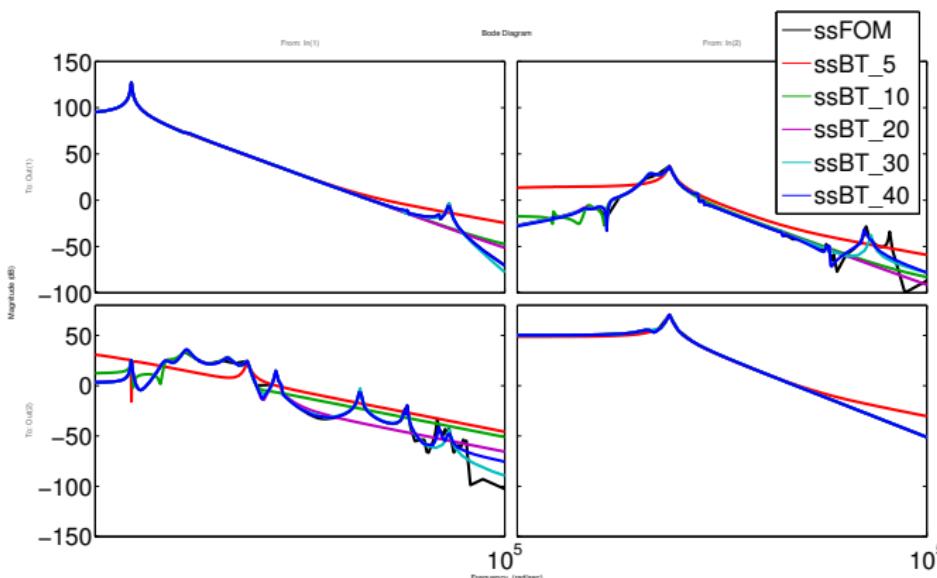
Bode plots associated with the HDM-based solution ( $N = 120$ ): Each column represents one input and each row represents a different output



## Application

## CD Player System (B. Salimbahrami and B. Lohmann, 2003)

Bode plots associated with the PROM-based (BT) solution: Each column represents one input and each row represents a different output



## └ Balanced POD Method

## └ Balanced POD Method

- The Balanced POD (BPOD) method generates two sets of snapshots: The standard POD solution snapshots; and the dual POD snapshots introduced below

$$\mathbf{S} = [(j\omega_1 \mathbf{I} - \mathbf{A})^{-1} \mathbf{B} \ \cdots \ (j\omega_{N_{\text{snap}}} \mathbf{I} - \mathbf{A})^{-1} \mathbf{B}]$$
$$\mathbf{S}_{\text{dual}} = [(-j\omega_1 \mathbf{I} - \mathbf{A}^*)^{-1} \mathbf{C}^* \ \cdots \ (-j\omega_{N_{\text{snap}}} \mathbf{I} - \mathbf{A}^*)^{-1} \mathbf{C}^*]$$

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- Next, BPOD computes right and left ROBs as follows

$$\begin{aligned}\mathbf{S}_{\text{dual}}^T \mathbf{S} &= \mathbf{U} \mathbf{\Sigma} \mathbf{Z}^T \quad (\text{SVD}) \\ \mathbf{V} &= \mathbf{S} \mathbf{Z}_k \mathbf{\Sigma}_k^{-1/2} \\ \mathbf{W} &= \mathbf{S}_{\text{dual}} \mathbf{U}_k \mathbf{\Sigma}_k^{-1/2}\end{aligned}$$

where the subscript  $k$  designates the first  $k$  terms of the singular value decomposition

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where the subscript  $k$  designates the first  $k$  terms of the singular value decomposition

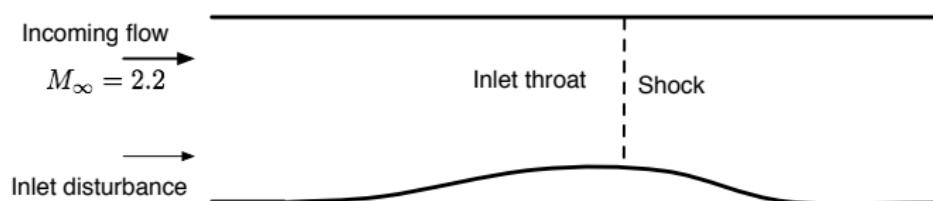
- If no truncation is performed, the result is equivalent to two-sided moment matching at  $s_i \in \{\omega_1, \dots, \omega_{N_{\text{snap}}}\}$  (see later)

## └ Balanced POD Method

## └ BT and POD in the Time Domain

- The POD method in the time domain is based solely on the reachability concept
- However, the BPOD method
  - adds the notion of observability in the construction of a PROM
  - is tractable for very large-scale systems
  - provides an approximation to the BT method
  - does not guarantee in general the stability of the resulting PROM

- Supersonic Inlet Problem (part of the Oberwolfach Model Reduction Benchmark Collection repository)



$$\begin{aligned}\mathbf{E} \frac{d\mathbf{w}}{dt}(t) &= \mathbf{A}\mathbf{w}(t) + \mathbf{B}\mathbf{u}(t) \\ \mathbf{y}(t) &= \mathbf{C}\mathbf{w}(t)\end{aligned}$$

- $N = 11\,370$  (2D Euler equations)
- $in = 1$  input (density disturbance of the inlet flow)
- $q = 1$  output (average Mach number at the diffuser throat)

└ Balanced POD Method

└ Application

- PMOR in the frequency domain using
  - POD
  - BPOD
- In both cases, same frequency sampling for the computation of solution snapshots

## └ Balanced POD Method

## └ Application

- PMOR in the frequency domain using
  - POD
  - BPOD
- In both cases, same frequency sampling for the computation of solution snapshots
- Plot of the magnitude of the relative error in the transfer function (within the sampled frequency interval) as a function of the dimension  $k$  of the constructed PROM

